## DIGITAL HILBERT TRANSFORMERS FOR FPGA-BASED PHASE-LOCKED LOOPS

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#### ABSTRACT

The phase detector is a main building block in phaselocked loop (PLL) applications. FPGAs permit the realtime implementation of the CORDIC algorithm which offers an efficient solution for an accurate phase detection, provided that the signal is available as an analytic signal. Different architectures for generating analytic signals by approximating the Hilbert transform were analyzed. Thereby, the focus has been on the demands based on the PLL application and the efficient implementation on FPGAs. Two methods were implemented using either FIR or complex filters. The FIR method results in a remaining phase error that has a zero mean value in time domain. An efficient IIR low-pass structure is proposed to suppress this phase error. The complex filters were implemented using a novel method based on complex, multiplier-less frequency sampling filters. Structures with different complexities are presented. A better result was achieved compared to a standard IIR filter design.

#### 1. INTRODUCTION

Recent advances in the technology of field programmable gate arrays (FPGA) enable the implementation of all-digital phase-locked loops (ADPLL) working directly in the high frequency range, unlike traditional micoprocessor-based approches. An offset-local oscillator (Offset-LO) PLL [1] was needed within a phase measurement system for the heavyion synchrotron (SIS18) accelerator at GSI. This system is used to determine the phase difference of RF cavities in relation to other cavities [2] or the beamphase [3] to compensate for deviations in a closed-loop control.

PLLs are widely used in communication systems. Many specialized integrated circuits exist that suit most of these applications. Due to the requirements of this Offset-LO, an FPGA-based implementation was realized instead, which particularly covers the relatively wide frequency range ( $f_{RF}$ = 0.8 – 5.4 MHz) and provides a phase accuracy better than 5° during the accelerating frequency ramp ( $f_{RF,max}$  = 68 MHz/s).

The phase detector is a key component in a PLL. Most VLSI implementations use phase detection methods, such as flip-flop based phase frequency detectors or time-to-digital converters, that are not feasible on FPGAs with an accuptable phase accuracy. A sophisticated way in detecting the phase is provided by the coordinate rotation digital computer (CORDIC) algorithm which was presented by Volder in 1959 and was enhanced by Walther in 1971 [4]. An existing VHDL implementation of the algorithm from Guntoro [5] could be readily used. The CORDIC algorithm implies that the input is a (complex) analytic signal, which consists of in-phase (real part) and quadrature component (imaginary part). The connection between in-phase (I) and quadrature component (Q) is given by the Hilbert transform which will be described in the next section.

Different methods exist for generating the analytic signal from a real valued signal. They are described and compared in Section 3. The implementation of a Hilbert transform with a single real FIR filter is described in Section 4. Eventhough modern FPGAs include hardware multipliers, these are still a limited resource in affordable devices. A novel method for designing complex multiplier-less frequency sampling filters as Hilbert transformers is presented in Section 5.

### 2. PHASE DETECTION WITH ANALYTIC SIGNALS

A high resolution phase detector can be realized by using the concept of the analytic signal. An analytic signal can formally be generated from a real signal by using the Hilbert transform. The discrete Hilbert transform can be formulated as a corrodution with a frequency response of

$$H_H(e^{j\Omega}) = \begin{cases} -j & \text{for } 0 < \Omega < \pi \\ j & \text{for } -\pi < \Omega < 0 \\ 0 & \text{for } \Omega = 0 \end{cases}$$
 (1)

with  $\Omega = 2\pi f/fs$ , where  $f_s$  is the sampling frequency in Hz and  $j = \sqrt{-1}$ . The absolute amplitude is unity for all frequencies not equal to zero. The phase is  $-90^\circ$  for all positive, and  $+90^\circ$  for all negative frequencies. The inverse Fourier transform results in the impulse response

$$h_H(n) = \frac{1 - \cos(\pi n)}{\pi n} = \begin{cases} 2/(\pi n) & \text{for odd } n \\ 0 & \text{for even } n. \end{cases}$$
 (2)

# <u>Digital Hilbert Transformers For Fpga Based Phase</u> <u>Locked</u>

Jonathan P. Skroch

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Science Abstracts ,1995 Electrical & Electronics Abstracts ,1997  $\square$  ,2006 Digital Hilbert Transformers Composed of Identical Allpass Subfilters ,1998

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